

Simulation of Three Phase Voltage Controlled Soft Switching Start of Induction Motor Drive

Bibhu Prasad Ganthia, Lipsa Ray, Priyanka Senapati, Subrat Kumar Barik



Abstract: In this paper a three phase voltage regulator (thyristor based) with RL load and induction motor load is simulated for fixed firing angle first. As was seen from the results obtained that the induction motor response is highly impaired with this fixed firing angle starting. The transient response as well as the steady state response is highly oscillatory in nature. Then a suitable firing scheme was developed to vary the firing angles of each thyristor in reference to the zero crossing of the respective phase voltages. The control circuit was studied for R-L load first and then the induction motor was simulated with this firing angle control scheme. The basic objective of this work is to improve the transient response of the voltage regulator fed induction motor. One basic requirement for the induction motor to have an improved transient response is that the applied voltage to the motor must gradually increase. The same was achieved with the proposed control logic. Next the fault mode (short circuit and open circuit switch fault) analysis of the induction motor was taken up. Till now not much work has been done on this fault tolerant induction motor drive. Here a 2-phase close loop control was adopted to improve response of the induction motor in fault mode. We used both the voltage control loop as well as the current control loop to do so. Unfortunately not much can be done on this fault tolerant operation as the time was very short. Till the speed and torque pulsation during a short circuited switch fault was greatly improved by adopting this two phase control strategy.

Index Terms: Soft Starter, Voltage Control Loop, Current Control Loop, Induction Machine, 2-Phase Control Loop, Fault Tolerant Operation.

I. INTRODUCTION

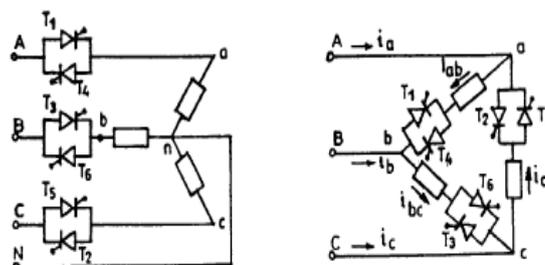
Poly-phase induction motors have been the workhorse (main prime movers) for industrial and manufacturing processes as well as numerous propulsion applications. The energization of such motors in these processes and applications can be achieved through the following ways:

- (1) Direct across-the-line starting
- (2) Soft-starting
- (3) Adjustable-speed drive (ASD) control

It is well known that direct across-the-line starting of induction motors, which offers absolutely no control capability, is characterized by high starting inrush currents and high starting torque pulsations. In certain applications such as conveyor belt drives, these high starting torque pulsations may result in belt slippage, which consequently may lead to undesirable damage to motor-load systems. Frequent direct across-the-line motor starting may introduce significant electrical and thermal stresses on motor bearings and winding components including insulation, heating in motor windings, as well as mechanical stresses on motor cages, shafts and load couplings. The adverse effects due to such stresses on motors may result in undesirable consequences such as squirrel-cage bar breakages, stator winding damage, and inter-turn short-circuit faults, which may lead to catastrophic failures in motors. The losses during starting or say during no-load make a significant contribution to poor efficiency of the induction motor. More to the cause the transient response in case of adjustable speed drives will be critical concern in case the motor is started by conventional starter. Accordingly, reduced voltage starters, or the so-called soft starters, are often employed as effective means to reduce high starting currents and torque pulsations through use of thyristor based voltage control both during starting as well as in adjustable speed drives. While the basic concept is simple, the problem of optimizing the performance of the so called soft starter is a complex task. Also the selection of the control variable and the control law is pretty important in order to improve the motor performance under various conditions.

II. THREE PHASE VOLTAGE CONTROLLER

The three phase loads that is to be fed from the regulator may be connected in different manner. The different connections are shown below:



Manuscript published on 30 September 2019

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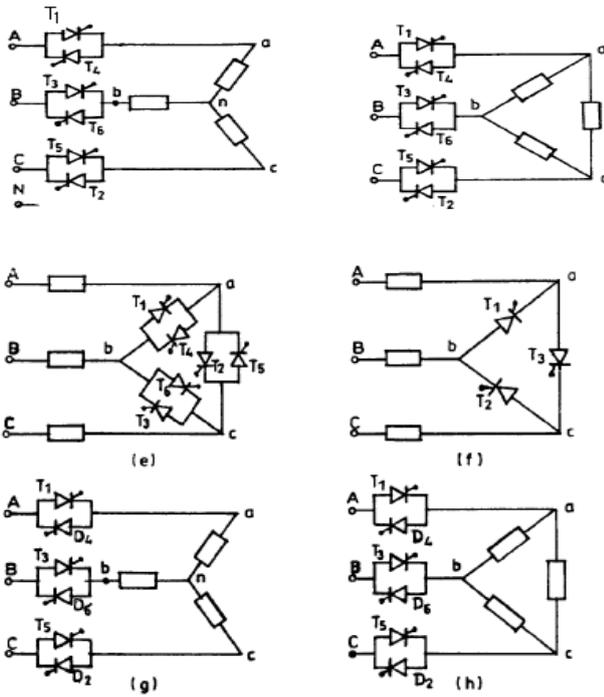


Figure.1 Different configuration three phase voltage regulator configuration [10]

Here we have considered the three phase star connected RL load with isolated neutral.

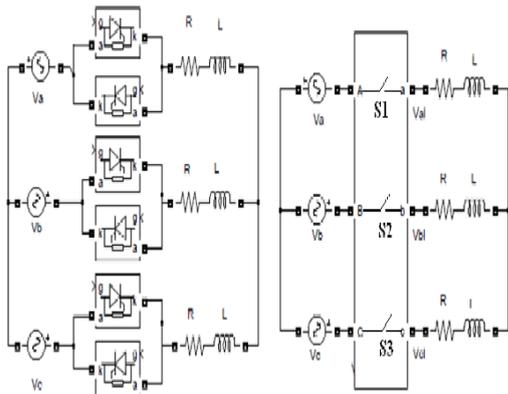


Figure.2 Three phase voltage regulator with RL load

The voltage regulator with RL load (star connected) is shown in figure (2.2). Let s_1 , s_2 and s_3 represents the three thyristor switches such that:

$$s_1 = \begin{cases} 1, & \text{when the phase a is conducting} \\ 0, & \text{when phase is not conducting} \end{cases}$$

$$s_2 = \begin{cases} 1, & \text{when the phase b is conducting} \\ 0, & \text{when phase is not conducting} \end{cases}$$

$$s_3 = \begin{cases} 1, & \text{when the phase c is conducting} \\ 0, & \text{when phase is not conducting} \end{cases}$$

Now at any moment any two phases or three phases are conducting. For this we define a new variable $s=1$ for any two or three phase conducting and is equal to zero for no phase conducting.

2.1.1. Voltage equations for different conduction modes:

We assume there is no mutual coupling between the phases, the voltage expressions for different modes are as follows-

Mode 1: when all three phases are conducting.

In this case $s_1=1, s_2=1, s_3=1$ and $s=1$. Then from the fig. it is clear that-

$$V_{al} = V_a; \quad V_{bl} = V_b; \quad \text{and} \quad V_{cl} = V_c$$

Mode 2: when two phases are conducting (phase a and phase b). That is $s_1=1, s_2=1, s_3=0$ and $s=1$ then from the fig.2.2

$$V_{al} = \frac{1}{2}V_{ab}; \quad V_{bl} = \frac{1}{2}V_{ba}; \quad V_{cl} = 0;$$

Mode 3: when two phases are conducting (phase a and phase c). That is $s_1=1, s_2=0, s_3=1$ and $s=1$ then from the fig.2.2

$$V_{al} = \frac{1}{2}V_{ac}; \quad V_{bl} = 0; \quad V_{cl} = \frac{1}{2}V_{ca};$$

Mode 4: when two phases are conducting (phase b and phase c). That is $s_1=0, s_2=1, s_3=1$ and $s=1$ then from the fig.2.2

$$V_{al} = 0; \quad V_{bl} = \frac{1}{2}V_{bc}; \quad V_{cl} = \frac{1}{2}V_{cb};$$

Mode 5: when two phases or all three phases are not conducting $s_1=0, s_2=0, s_3=0$ and $s=0$, so from the fig.2.2

$$V_{al} = 0; \quad V_{bl} = 0; \quad V_{cl} = 0;$$

So the generalized load voltage for all states can expressed as follows:

$$\left. \begin{aligned} V_{al} &= \frac{s_1}{1+s_2+s_3} ((s_2 + s_3)V_a - s_2V_b - s_3V_c) \\ V_{bl} &= \frac{s_2}{1+s_1+s_3} ((s_1 + s_3)V_b - s_1V_a - s_3V_c) \\ V_{cl} &= \frac{s_3}{1+s_1+s_2} ((s_2 + s_1)V_c - s_1V_a - s_2V_b) \end{aligned} \right\} \quad (2.2)$$

Where V_a, V_b, V_c are the instantaneous supply voltages.

III. INDUCTION MOTOR SOFT STARTER

A soft starter is normally used to reduce the starting voltage automatically so that the magnetizing inrush reduces. This may be done by using open loop control or close loop control. The open loop control delays the application of the firing pulses to the thyristors and with a set time delay it produces the pulses. Here we go for close control of the soft starter. As mentioned earlier in the previous chapter, soft starter technology is widely employed as effective and low-cost means, as compared to modern PWM adjustable-speed drives, to reduce high starting currents and torque pulsations of medium voltage and large ac motors in numerous critical industrial, manufacturing, and transportation applications through use of thyristor-based voltage control [11][15]. The circuit topology and its control scheme are simple and easy to implement, which will be described in the following subsection. In an induction motor, the set of differential equations in dq representation that govern the instantaneous relationship between the voltages and currents in the synchronously rotating frame of reference can be expressed as follows[21]:

$$\begin{aligned}
 v_{qs}^e &= r_s i_{qs}^e + \frac{d\lambda_{qs}^e}{dt} + \omega_e \lambda_{ds}^e \\
 v_{ds}^e &= r_s i_{ds}^e + \frac{d\lambda_{ds}^e}{dt} - (\omega_e - \lambda_{qs}^e) \\
 0 &= r_r i_{qr}^e + \frac{d\lambda_{qr}^e}{dt} + (\omega_e - \omega_r) \lambda_{dr}^e \\
 0 &= r_r i_{dr}^e + \frac{d\lambda_{dr}^e}{dt} - (\omega_e - \omega_r) \lambda_{qr}^e
 \end{aligned}
 \tag{3.28}$$

Where,

$$\begin{aligned}
 \lambda_{qs}^e &= L_s i_{qs}^e + L_m i_{qr}^e \\
 \lambda_{ds}^e &= L_s i_{ds}^e + L_m i_{dr}^e \\
 \lambda_{qr}^e &= L_r i_{qr}^e + L_m i_{qs}^e \\
 \lambda_{dr}^e &= L_r i_{dr}^e + L_m i_{ds}^e
 \end{aligned}
 \tag{3.29}$$

Here $L_s = L_{ls} + L_m$ and $L_r = L_{lr} + L_m$, where, L_{ls} the stator leakage inductance, L_{lr} is the referred rotor leakage inductance, and L_m is the magnetizing inductance in the T- equivalent circuit of an induction machine [22]. The first two equations of (3.28) correspond to the dq stator windings, while the last two equations of (3.28) correspond to the dq rotor windings. The equations of (3.29) represent the dq stator and rotor flux linkages. Meanwhile, the terms, ω_e and ω_r , represent the synchronous and rotor speed in radians/sec. subscript denotes synchronously rotating frame. In principle, reducing the impressed voltage upon the motor during starting reduces the starting current and torque pulsations. This is due to the fact that the starting torque or locked rotor torque is approximately proportional to the square of the starting current or locked rotor current, and consequently it is proportional to the square of the starting voltage [17]. Therefore, by properly adjusting the applied effective voltage during startup, the locked rotor torque and current can be reduced. The most common control strategy employed by the soft starter of Figure 2.1 is the open loop voltage control. Such control approach is widely adopted in commercially available soft starters, as well as soft starter designs reported in the literature [18]-[19]. The voltage control is implemented by adjusting either the delay angle, α , or the hold-off angle, γ , of the conduction cycle of the oncoming thyristor with respect to either the zero crossing of the supply voltage (α) or the zero crossing of the line current (γ), respectively, as shown in Figure 2.2. The thyristors are then selectively fired to conduct current in the appropriate phase, and naturally commutate off when the current reaches zero. The larger the delay angle, α , or the hold-off angle, γ , the larger the notch width in the applied motor voltage, which consequently reduces the effective or RMS value of such voltage impressed upon the motor. However, improper control of α or γ firing angles may result in relatively high starting torque and current oscillations. Therefore, optimum starting profiles of the α or γ firing angles have been extensively investigated to produce smooth starting torque and current profiles.

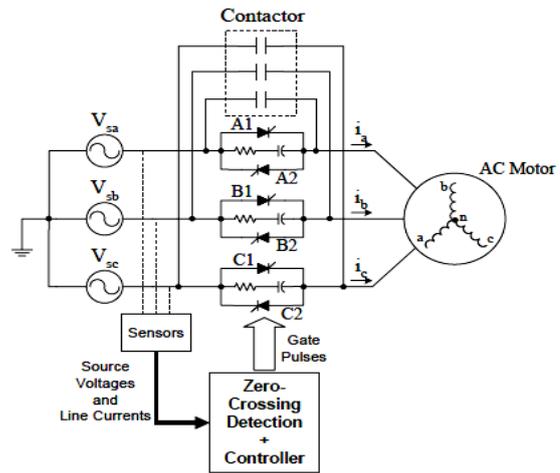
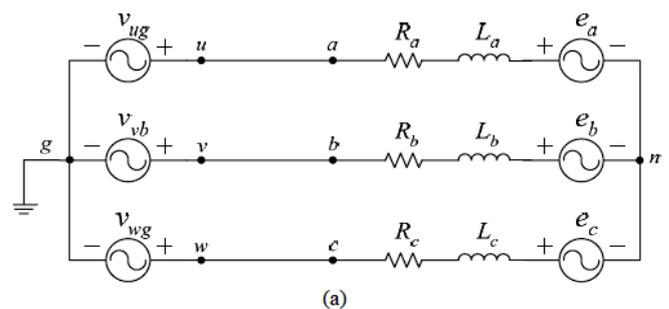


Figure.3 Conventional three-phase soft starter topology

In order to obtain good dynamic characteristics, a symmetrical-triggering firing pulse sequence is necessary. Due to the fact that the thyristor is a half-controlled device, its firing pulse must be synchronized with the ac mains. The six thyristors will be turned on with the proper sequence, and the firing signal is in phase with the zero-crossing of the supply voltage. Since there is an inherent delay associated with a thyristor turn-off, the input current of the motor is not continuous, which results in time harmonics in the machine air gap flux. These generated harmonics are generally odd order harmonic terms. In the commercially available soft starters, this firing angle profile is preset by the user based on the initial setting of the locked-rotor torque (LRT) and the ramp time, a simple illustration of which is shown in Figure 3.3. The LRT is the initial starting torque that is required to accelerate the motor during starting, and the ramp time is the time it takes for the voltage to go from the initial voltage value at the LRT setting to the maximum full voltage that is being applied to the motor. By adjusting the ramp time, the acceleration time of the motor can be controlled. Accordingly, the firing angle, which controls the amount of applied motor voltage, is reduced gradually depending on the ramp time during the period of starting until the motor has reached its full speed and rated current, whereupon the contactors are closed to bypass the thyristors. Evidently, such predetermined α or γ firing angle control offers very limited control flexibility. It should be noted from Figure 3.2 that the phase angle, ϕ , is related to the delay angle, α , and the hold-off angle, γ , by the following relation:

$$\phi = \alpha - \gamma$$



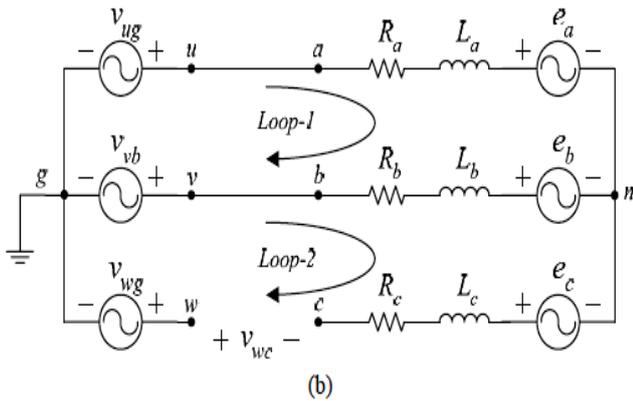


Figure.4 (a) 3-phase conduction mode. (b) 2-phase conduction mode.

IV. FAULT MODE ANALYSIS

The earlier formulations of the closed-form analytical expressions of the motor phase voltages and currents are employed herein to investigate the impact of failure modes in soft starters on the motor transient performance. Two distinct types of failure modes considered in this work are: (1) short-circuit SCR fault (see Figure 4.1(a)), and (2) open-circuit SCR fault (see Figure 4.1(b)), occurring only in one phase of the soft starter. A short-circuit SCR fault can happen in situations such as loose wire in the circuit or breakdown in the snubber circuit. Conversely, an open-circuit SCR fault can occur due to malfunctions either in the gate driver or the pulse generator of the controller.

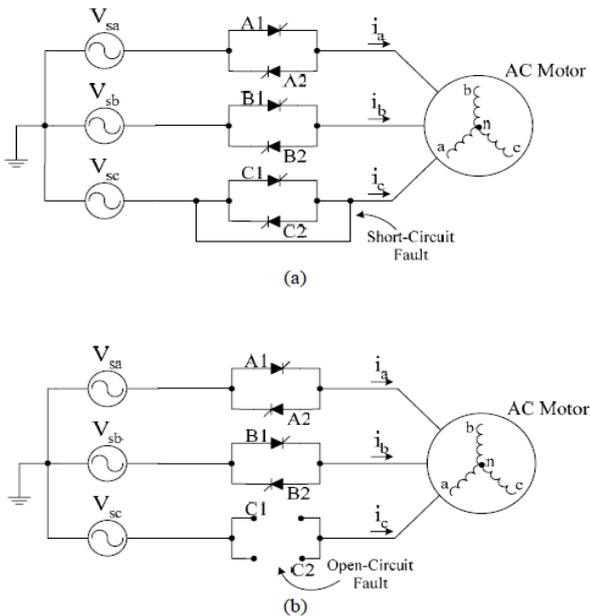


Figure.5 Failure modes in soft starter circuit (a) Short-circuit fault. (b) Open-circuit fault

V. SIMULINK MODEL AND RESULTS

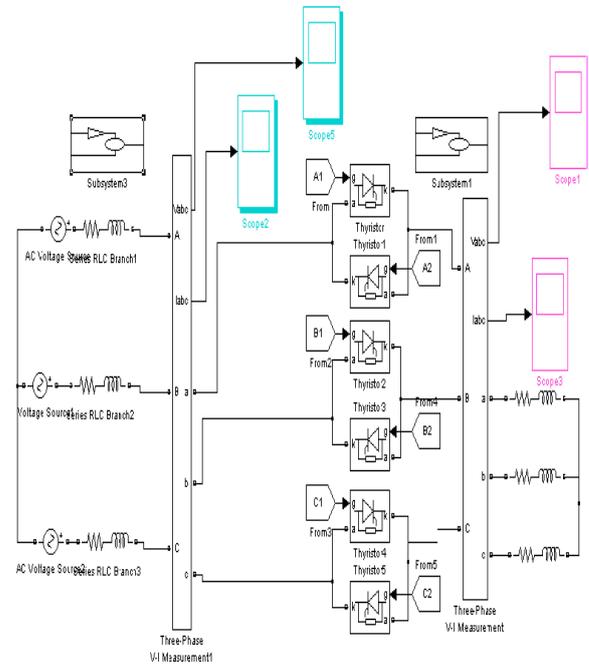


Figure.6 Simulink model of RL load

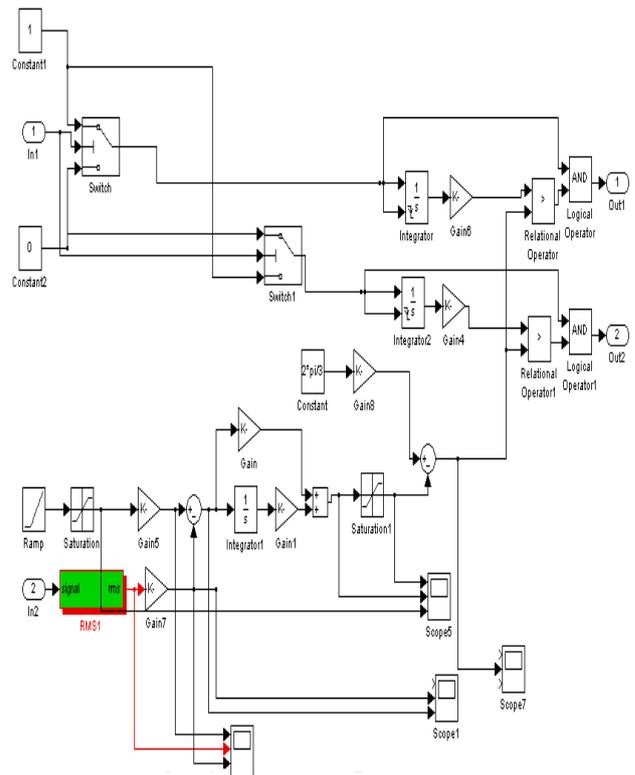


Figure.7 Firing circuit model for RL load

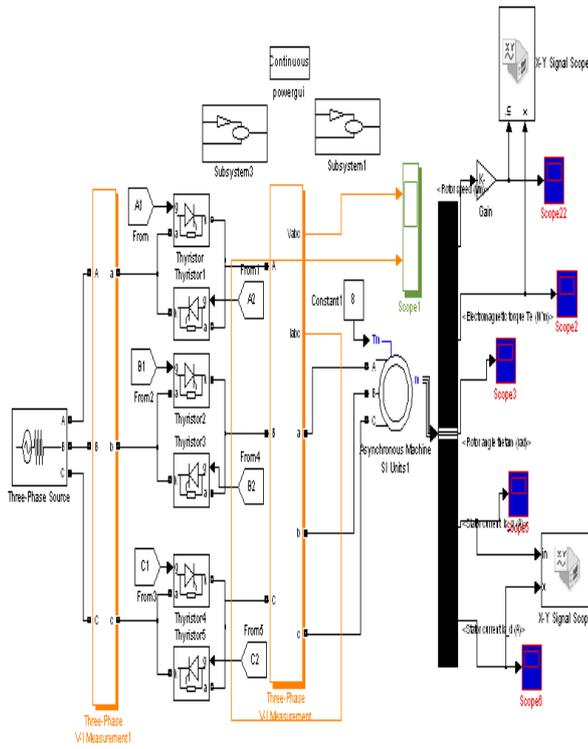


Figure.8 Induction motor model in healthy mode (soft starting)

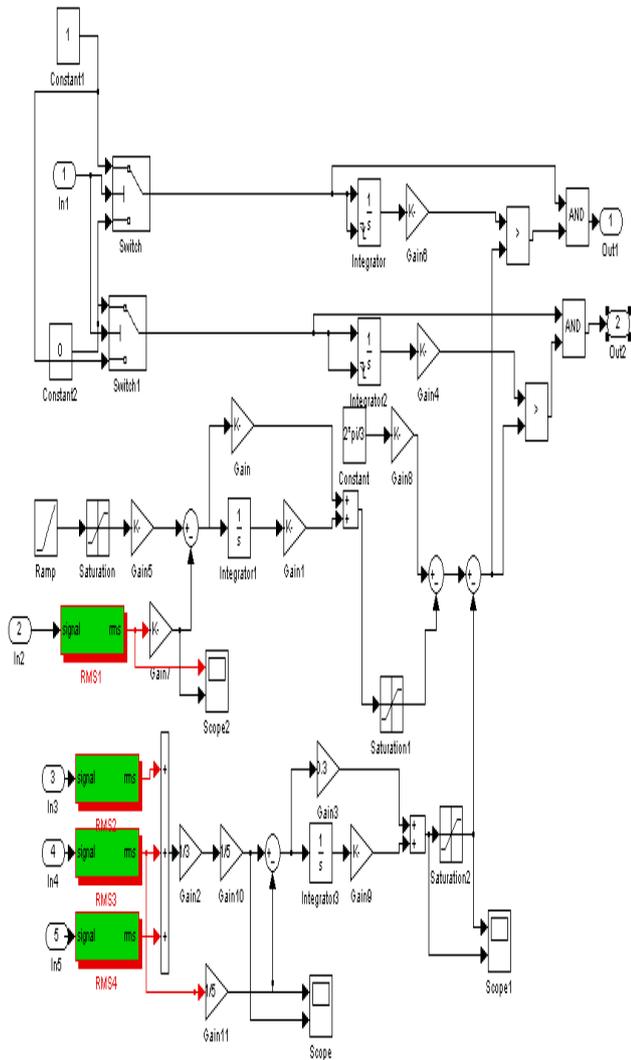


Figure.9 Proposed Control circuit model in fault mode (2-phase close loop control)

a. Simulation results for the RL load

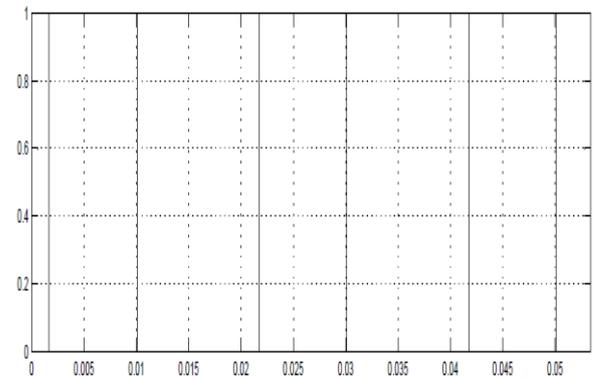
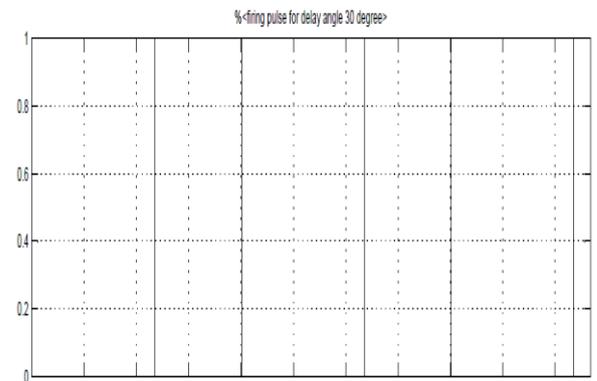
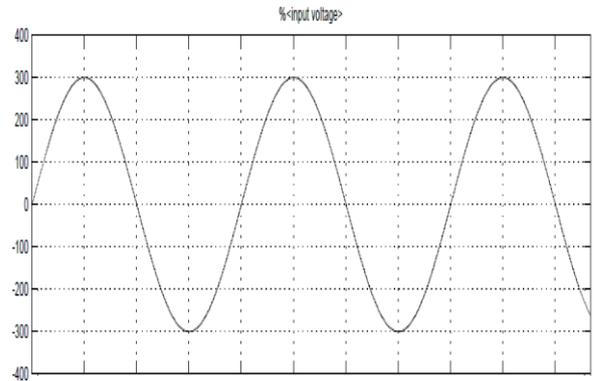
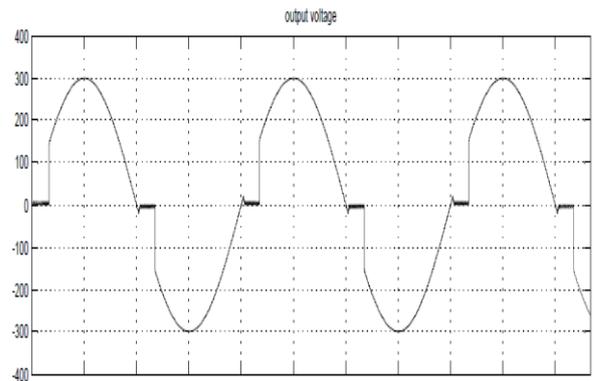


Figure.10 Voltage waveform with RL load for $\alpha = 30^\circ$

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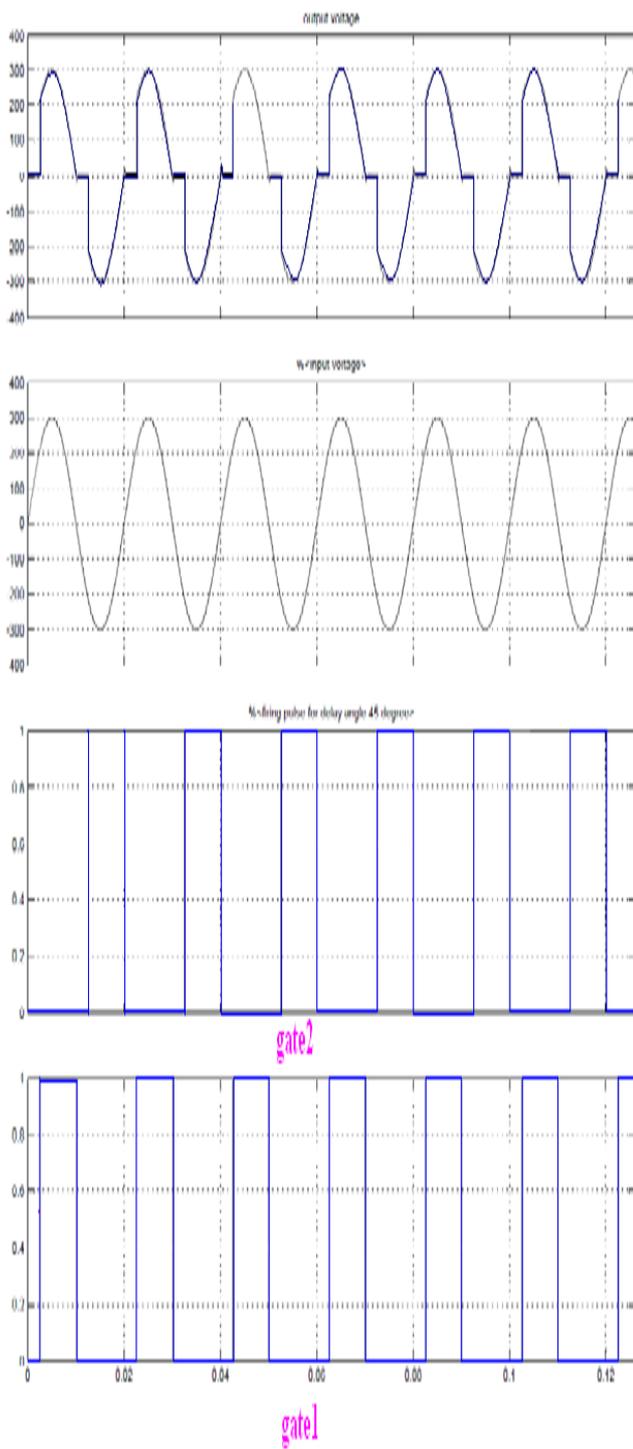


Figure.11 Voltage waveform with RL load for $\alpha = 45^\circ$

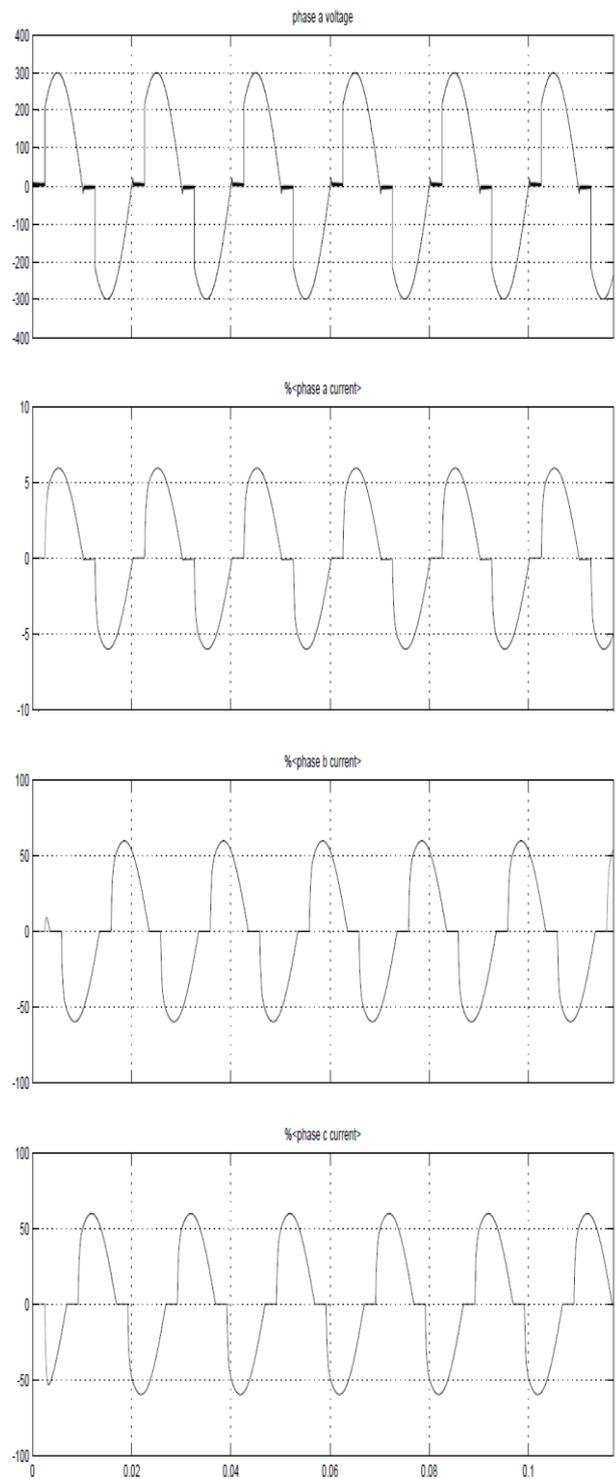


Figure.12 Current waveform with RL load for $\alpha = 45^\circ$

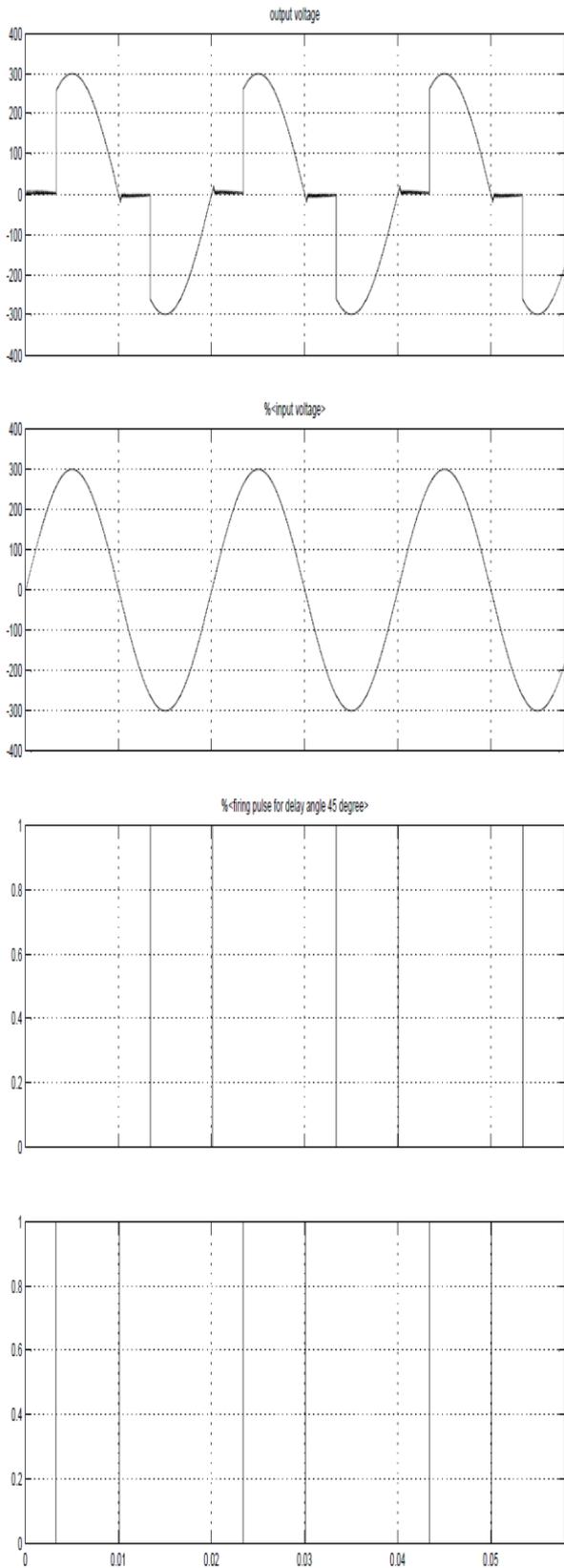


Figure.13 Voltage waveform with rl load for $\alpha = 60^{\circ}$

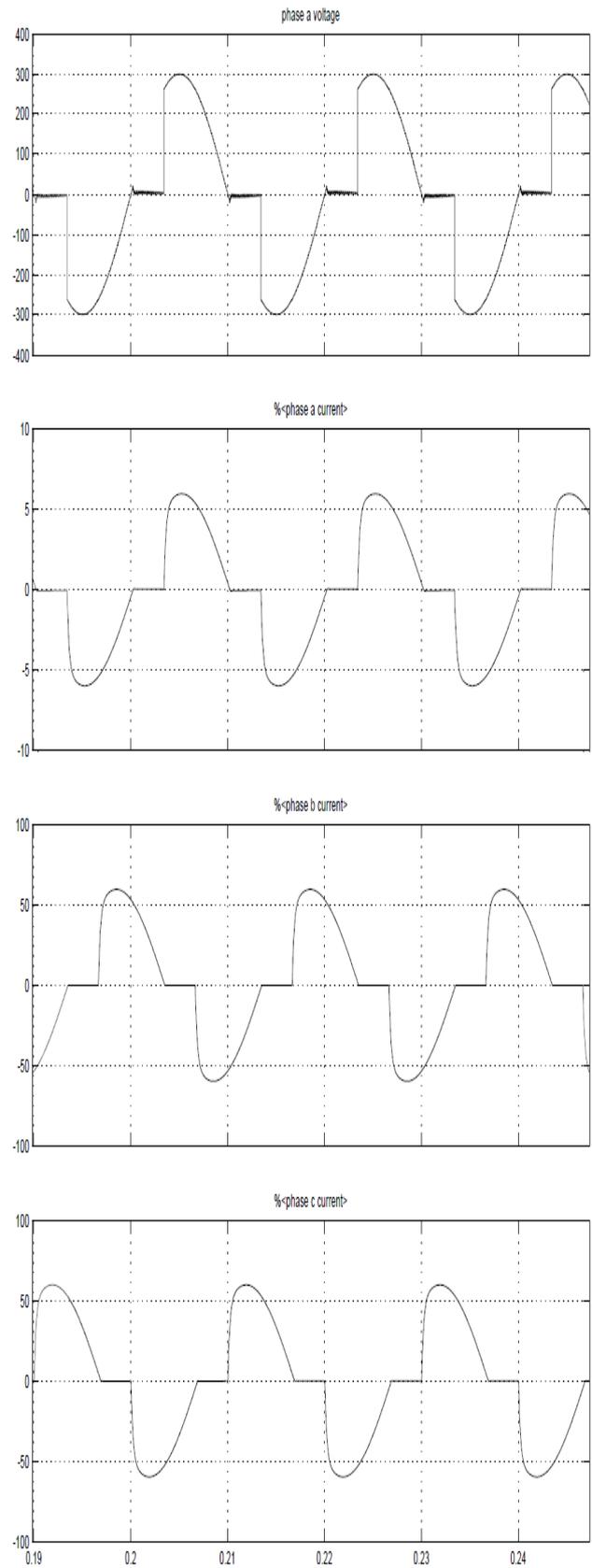


Figure.14 Current waveform with rl load for $\alpha = 60^{\circ}$

Simulation of Three Phase Voltage Controlled Soft Switching Start of Induction Motor Drive

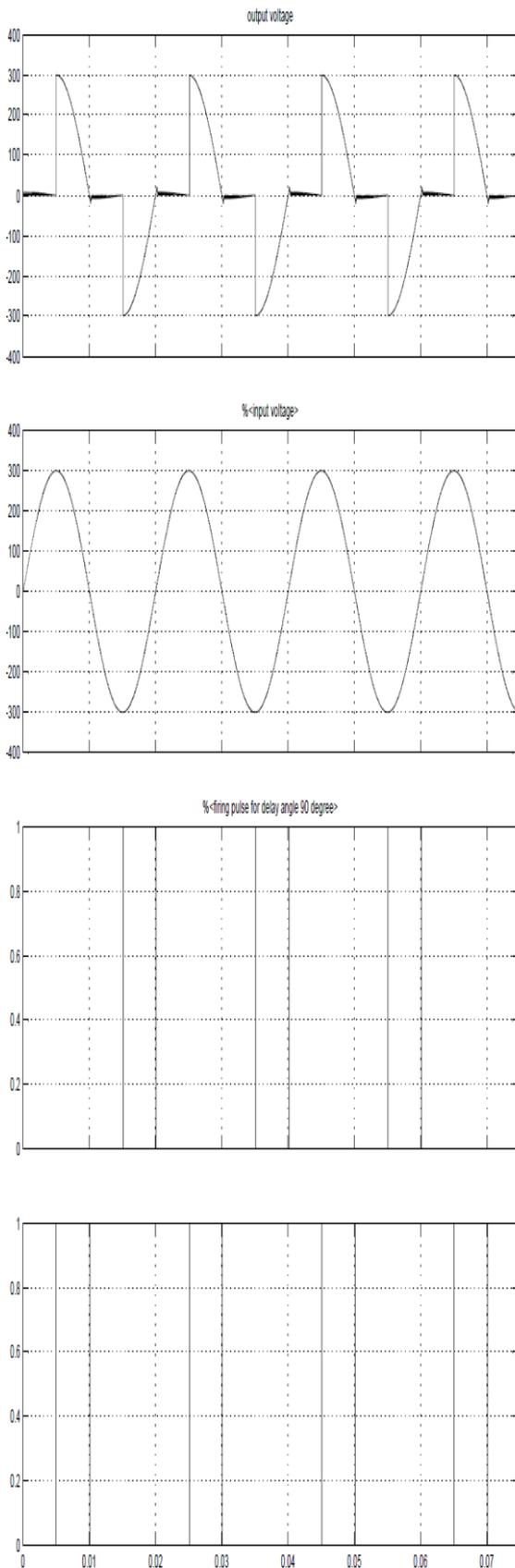


Figure.15 Voltage waveform with rl load for $\alpha = 90^\circ$

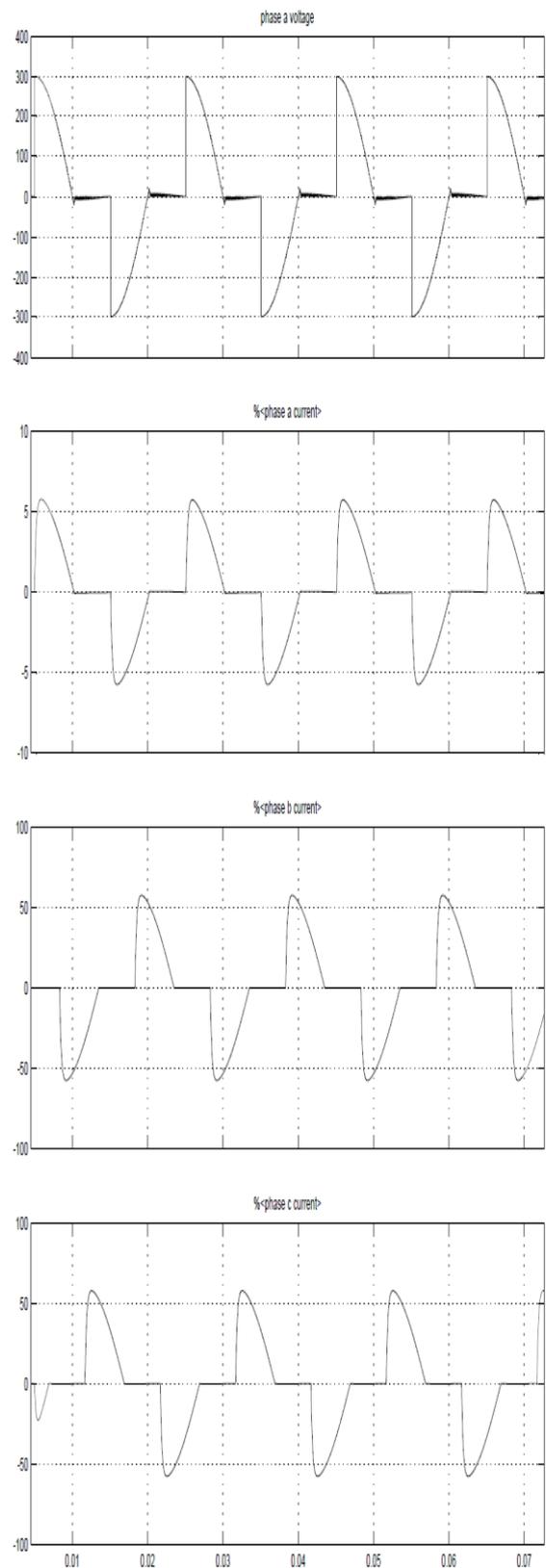


Figure.16 phase voltage and Current (a,b,c) waveform with rl load for $\alpha = 90^\circ$

Simulation result for R=30Ohm and L=50Mho

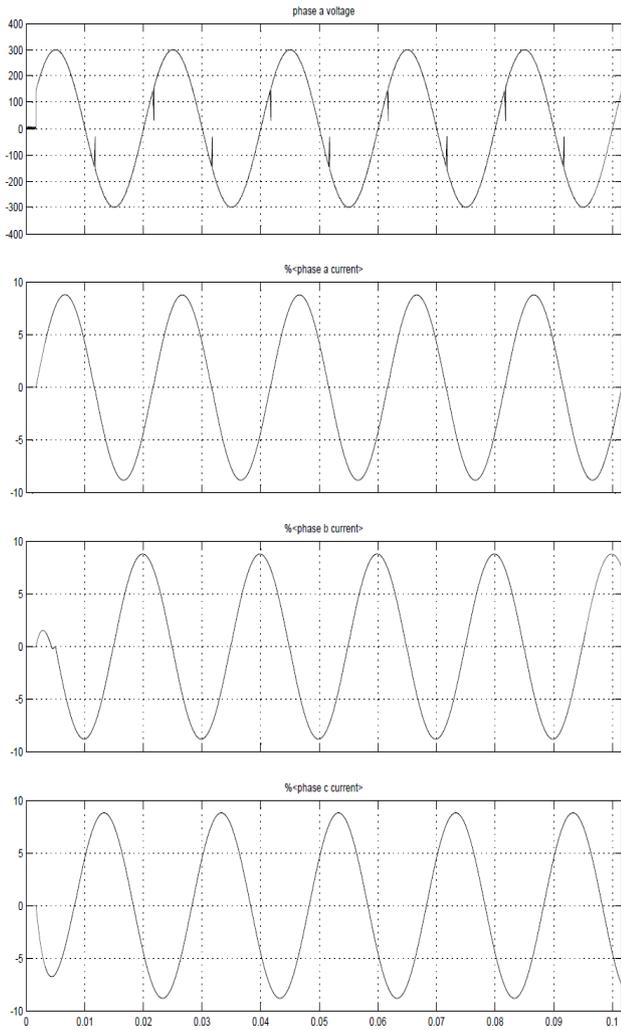


Figure.17 (a): Phase Voltage and current waveform with rl load for $\alpha = 30^\circ$

b. Simulation results for the 3-phase voltage regulator fed induction motor

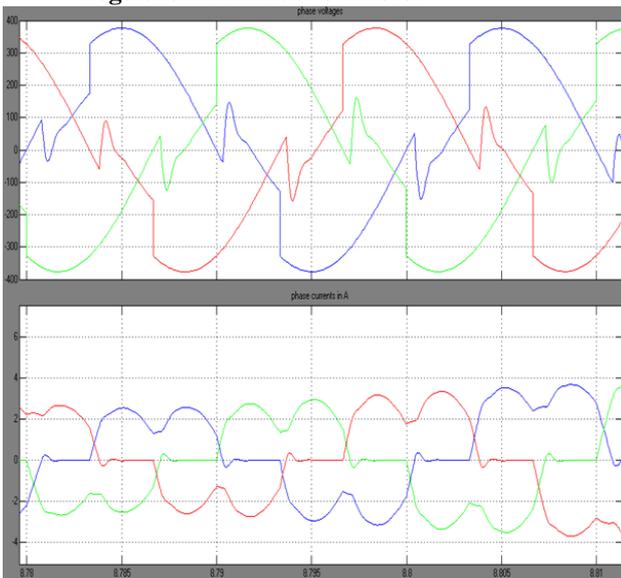


Figure.18 Phase voltage and current for delay angle 60°

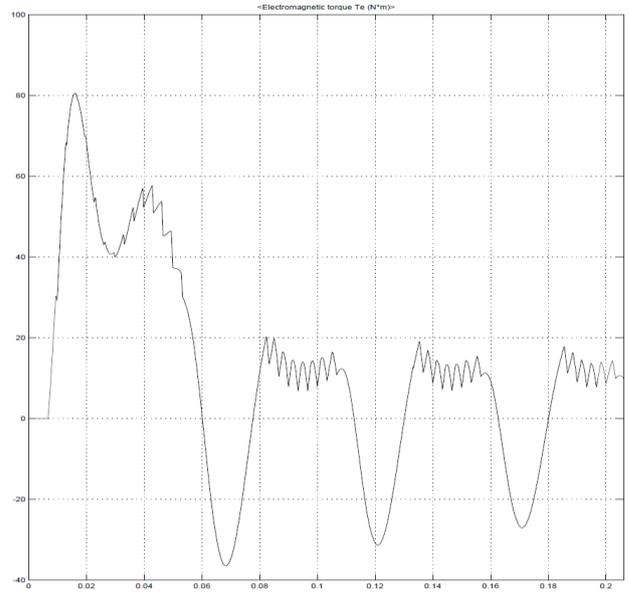


Figure.19 First few cycles of torque for delay angle 60°

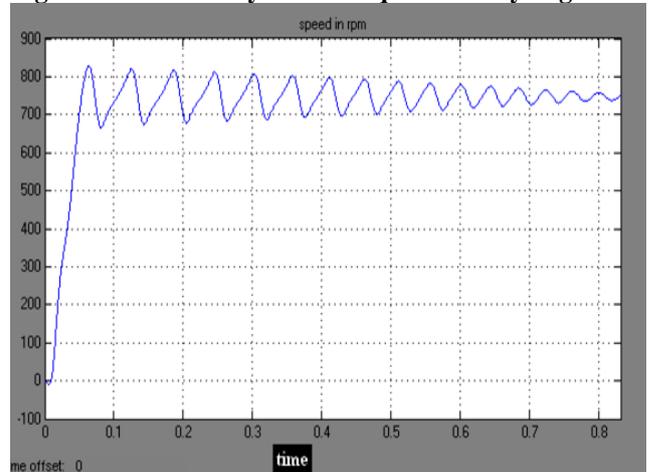


Figure.20 Speed versus time for delay angle 60°

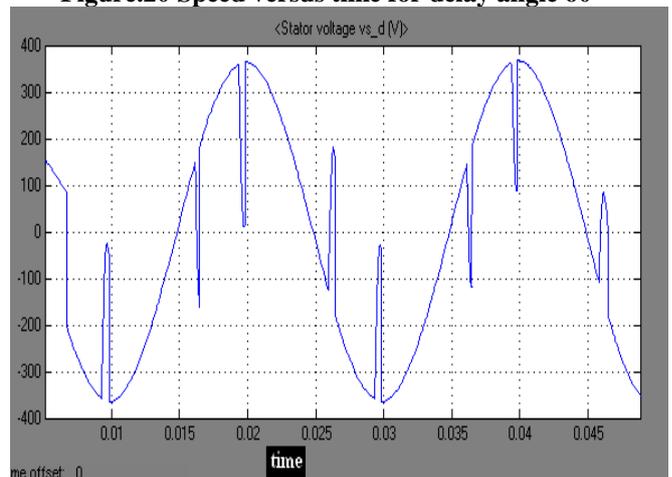


Figure.21 Phase voltage for delay angle 60° and 50% load torque

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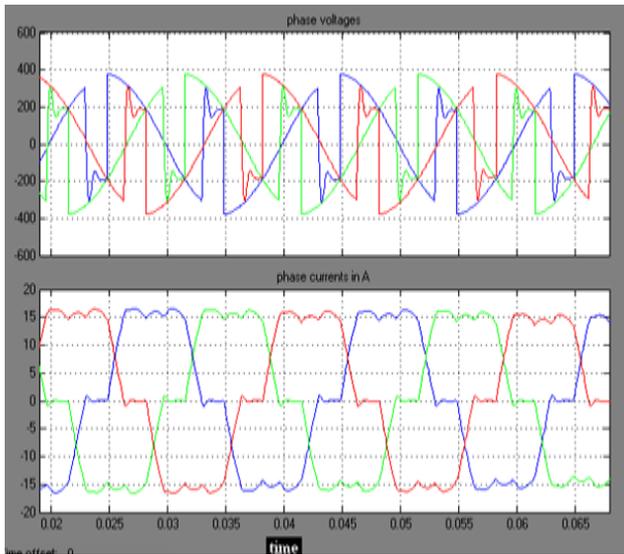


Figure.22 1st few cycles of phase current and phase voltages for delay angle 90° and 50% load torque

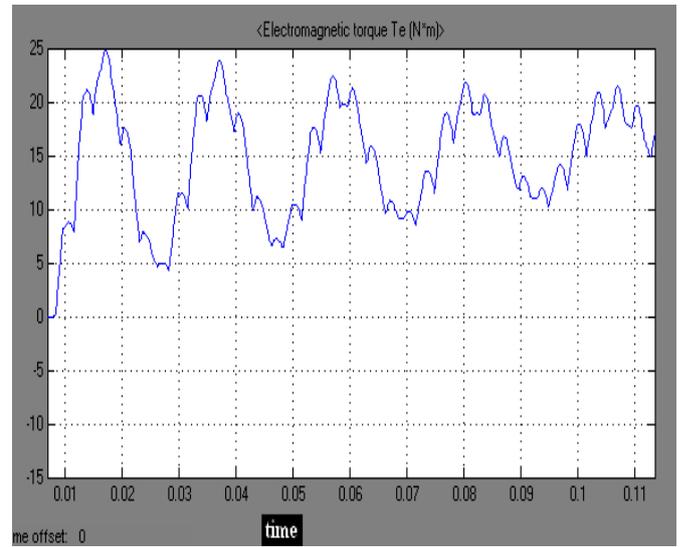


Figure.25 First few cycles of torque for delay angle 90° rated load

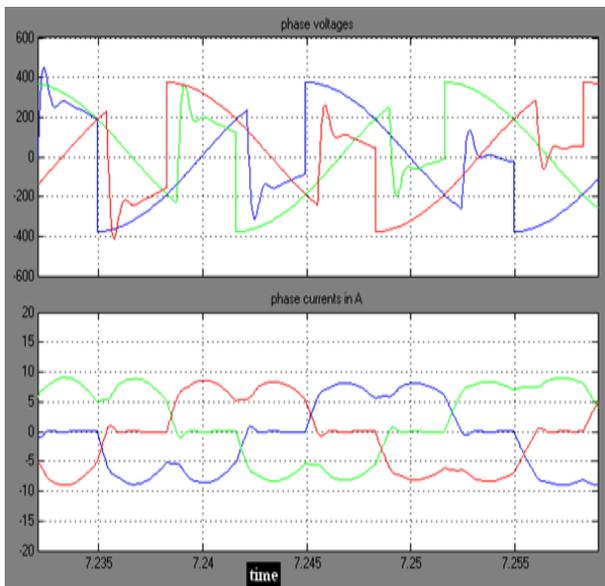


Figure.23 1st few cycles of phase current and phase voltages for delay angle 90° and 50% load torque

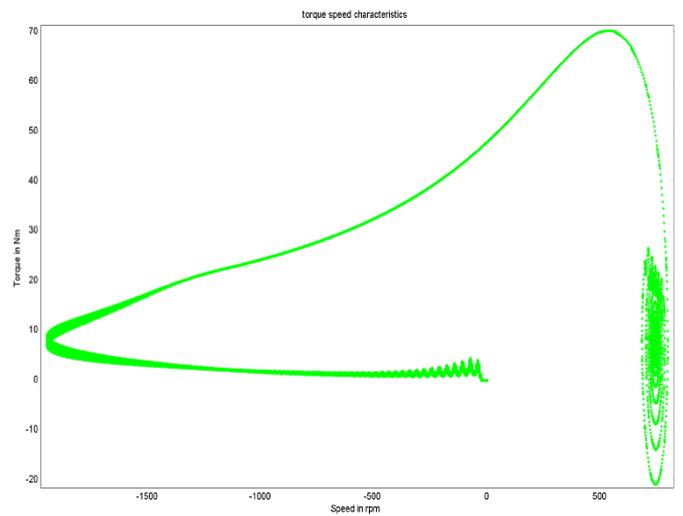


Figure.26 Torque versus speed characteristics of close loop controlled induction motor

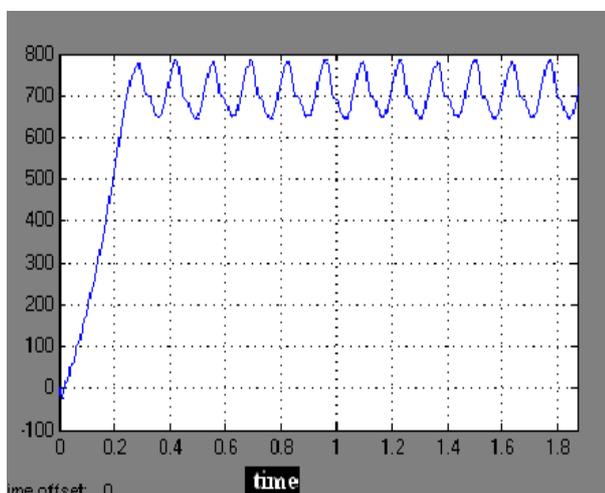


Figure.24 Speed versus time for delay angle 90°

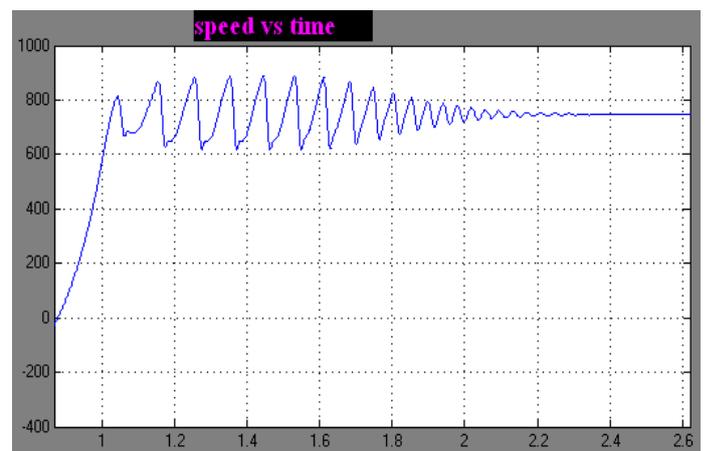


Figure.27 Speed versus time characteristics of close loop controlled induction motor

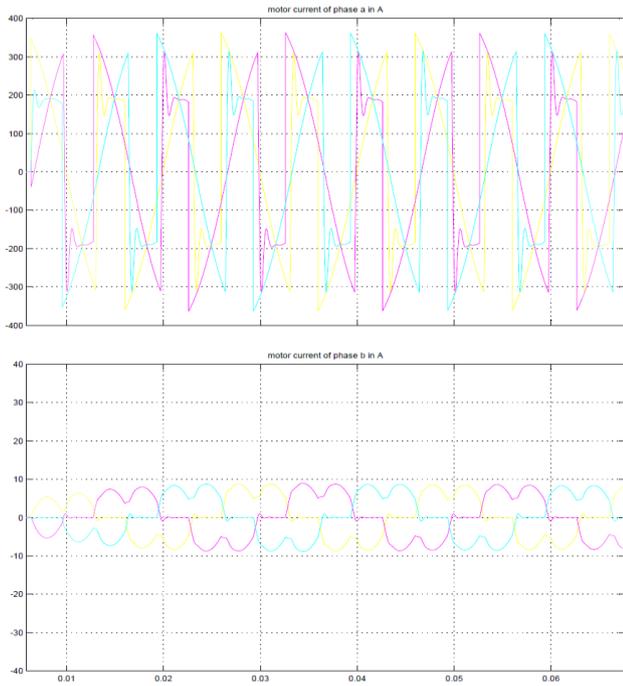


Figure.28 Transient voltage and current of close loop controlled induction motor

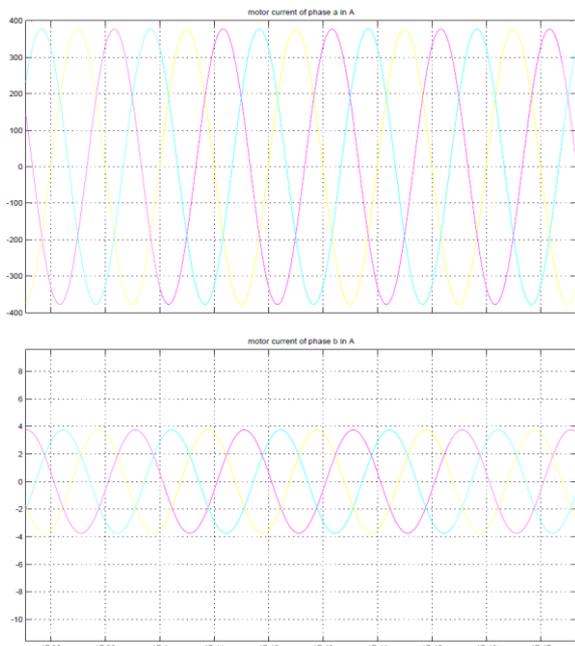


Figure.29 Steady state voltage and current of close loop controlled induction motor

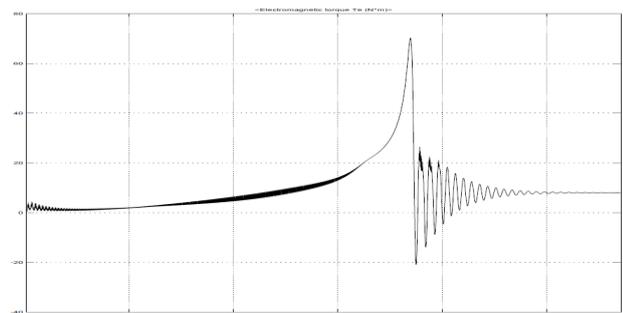


Figure.30 Torque versus time 1st few cycles of close loop controlled induction motor

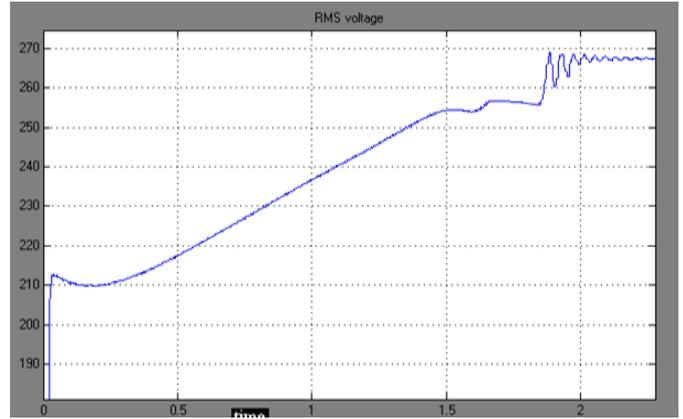


Figure.31 RMS output voltage (phase B) of close loop controlled induction motor

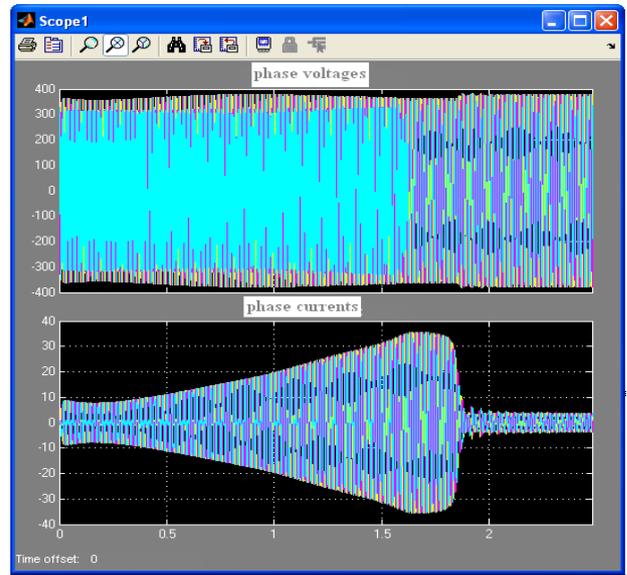


Figure.32 Phase voltages and phase currents of close loop controlled induction motor

Simulink Results of Fault Mode

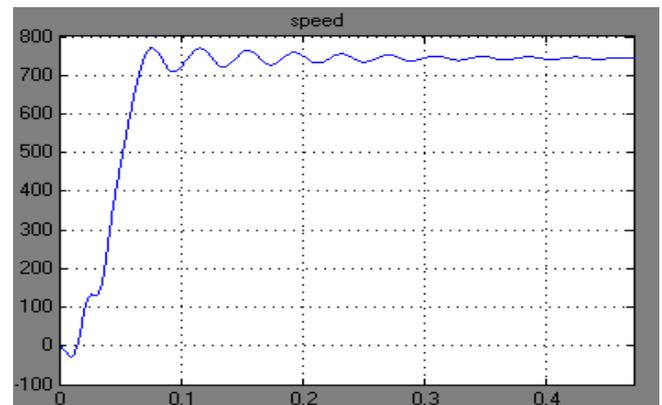


Figure.33 Speed versus time graph in case short circuit fault

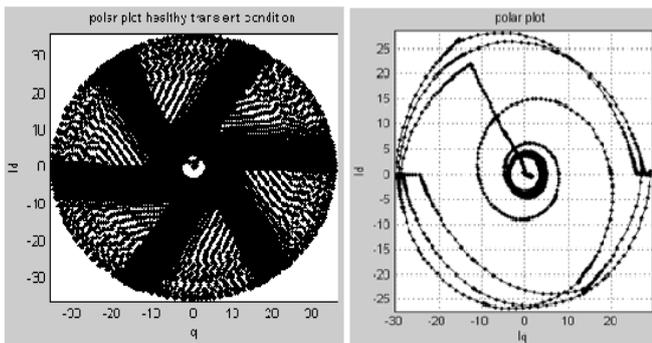


Figure.34 Polar plot (healthy condition) and Polar plot (short ckt)

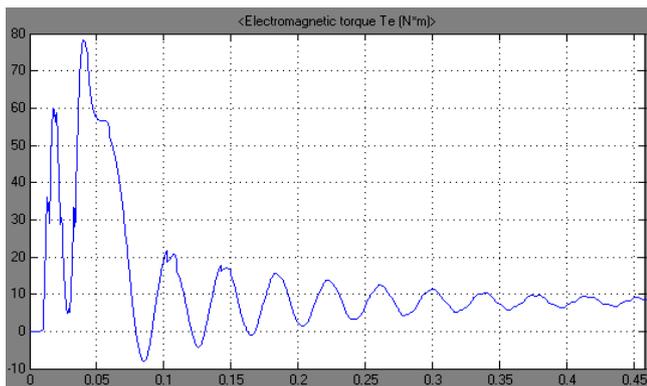


Figure.35 Torque versus time during a short circuit fault

VI. CONCLUSION

This paper is based on a simulation study of a three phase voltage controller with RL and induction motor load. Both the cases were also simulated for soft starting. The results obtained are pretty satisfactory. The initial magnetic inrush is greatly reduced with the help of a soft starter (thyristor voltage regulator based) along with the torque pulsation reduction. Also closed form derivation of the induction motor is conceived to analyze the transient performance during soft starting. Next the fault tolerant operation of the induction motor under various switch faults was studied. An analytical study was carried out but the simulation could not be carried satisfactorily because of shortage of time. But still the results obtained for the short circuit fault seems to be highly improved by this 2- phase close loop control method.

APPENDIX

Machine ratings

Rated Power 2 hp (1492 Watts)
 Rated Voltage (Line-Line) 460 Volts
 Rated Current 3.0 Amps
 Rated Frequency 50 Hz
 Rated Speed 1440 rpm
 Rated Torque 8.169 Nm
 Phase 3
 Number of Poles 4
 Stator Resistance, $R_s=3.850 \Omega$
 Rotor Resistance, $R_r= 2.574 \Omega$
 Stator Leakage Inductance, $L_{ls} = L 17.5594$ mH
 Rotor Leakage Inductance, $L_{lr} = 17.5594$ mH
 Magnetizing Inductance, $L_m = 0.372674$ H
 Moment of Inertia, $J 0.028$ kg.m²
 Load Coefficient, $k_L= 0.24493 \times 10$ Nm/ (m. rad/s)²

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