

Channel Estimation Techniques for OFDM and UPMC Systems for 5G Communications

Tipparti Anil Kumar, Sk Nilofer, Rajidi Sahithi

ABSTRACT—In this paper, pilot-assisted techniques for channel estimation (CE) are simulated for Universal Filtered Multi-Carrier (UPMC) modulation scheme. UPMC aims at replacing orthogonal frequency division multiplexing (OFDM) and improves performance and robustness in the case of time-frequency misalignment. These techniques efficiently support Internet of Things (IoT) and massive machine type communications (mMTC), which are identified as challenges for 5G wireless communication systems (WCS). Pilot-aided techniques are adopted and applied to OFDM and UPMC. Simulation results are supplemented to compare the performance of UPMC systems with conventional CP-OFDM systems.

Index terms:—Channel estimation; 5G; timing offset; synchronization; OFDM; UPMC;

I. INTRODUCTION

A complete system redesign is needed for next generation 5G WCS to support rapid growth in M2M communications and IoT [1]. In order to minimize the signaling overhead and battery and power consumption for low-end devices like simple sensor elements in 5G, they should be transmitted with synchronization conditions relaxed w.r.t time-frequency misalignments [2].

As OFDM is sensitive to these misalignments due to high spectral side-lobe level [3], and filter-bank based multi-carrier (FBMC) [4] is disadvantageous for communication in short uplink bursts (as required in possible application situations of 5G systems, for example low latency communication or energy-efficient mMTC, UPMC [5-8], which may be assumed as commonality of filtered OFDM and FBMC, is proposed for future 5G WCS. In this paper, the CE techniques are analyzed and compared OFDM and UPMC systems for 5G. Channel estimation techniques are discussed related to different channel models and different scenarios with different aspects. Simulation results are shown and performance comparison is made with respect to system parameters.

II. MULTI-CARRIER MODULATION TECHNIQUES

Multi-carrier systems advantageous over single carrier systems in terms of spectral efficiency [9], data rate,

immunity to frequency-selective and non-selective channels, and adaptive power allocation etc.,

2. a. Universal Filtered Multi-Carrier

The UPMC system model is shown in Fig.1

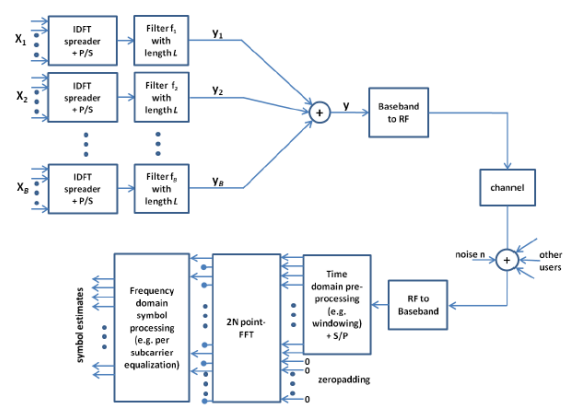


Fig.1: System model of UPMC.

The system bandwidth is sub-divided into number of B sub-bands. The every sub-band is represented as number of physical resource element in LTE system N_B , N is the total subcarriers. N-point IDFT is performed to transform signal from frequency to time domain. Symbols are modulated and zeros are padded in frequency domain to perform IDFT. The signal x_i is filtered by FIR-filter f_i of L. Therefore, sub band i output with symbol length $N+L-1$ (linear convolution of x_i and f_i) is given by

$$y_i(k) = x_i * f_i = \sum_{l=0}^{L-1} f_i(l)x_i(k-l), \quad k = 0, 1, \dots \quad (1)$$

3. Single-user channel estimation

If channel assisted ISI is negligible, then the channel estimates for given signal $X(k)$ for OFDM and UPMC at sub carrier k maybe expressed as eq. (1) and eq. (2) respectively [6]

$$\hat{H}(k) = \frac{Y_o(k)}{X(k)} = H(k) + \frac{N(k)}{X(k)} \quad (2)$$

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Dr. Tipparti Anil Kumar, Department of Electronics & Communication Engineering, CMR Institute of Technology, Kandlakoya (v), Medchal Road, Hyderabad – 501401. (tvakumar2000@yahoo.co.in)

Sk Nilofer, Department of Electronics & Communication Engineering, CMR Institute of Technology, Kandlakoya (v), Medchal Road, Hyderabad – 501401. (nilofershaik@gmail.com)

Rajidi Sahithi, Department of Electronics & Communication Engineering, CMR College of Engineering & Technology, Kandlakoya (v), Medchal Road, Hyderabad – 501401. (sahithiramu@gmail.com)

$$\hat{H}(k) = \frac{Y_U(k)}{X_i(k)F_i(k)} = H(k) + \frac{N(k)}{X_i(k)F_i(k)} \quad (3)$$

where $Y(k)$ is the received signal, $N(k)$ is AWGN and $F_i(k)$ is frequency response of FIR filter.

Therefore, the CE of UFMC systems differs in the equalization of shape of the filter. Mathematically, final channel estimates for single user systems with sliding window algorithm may be given as [7]

$$\hat{H}_{SW}(k, l) = \frac{1}{L_{SW}} \left(\hat{H}(k, l) + \sum_{n=1}^{(L_{SW}-1)} \left(\hat{H}(k-n, l) + \hat{H}(k+n, l) \right) \right) \quad (4)$$

where L_{SW} is size of the selection window.

In simulations, flat-fading channel is considered, simulation parameters FFT size 1024, Filter/ CP length: 74/ 73, PRB size: 12, No. of PRBs: 10, Modulation: QPSK and speed of the user is 50 kmh. Symbol error rate (SER) vs Eb/ No are simulated in Fig. 1 and 2. From simulation results it is observed that SER reduces as the sliding window size increases. According to the obtained SER values, size of sliding window of 7 or 9 may be selected for flat-fading channel model.

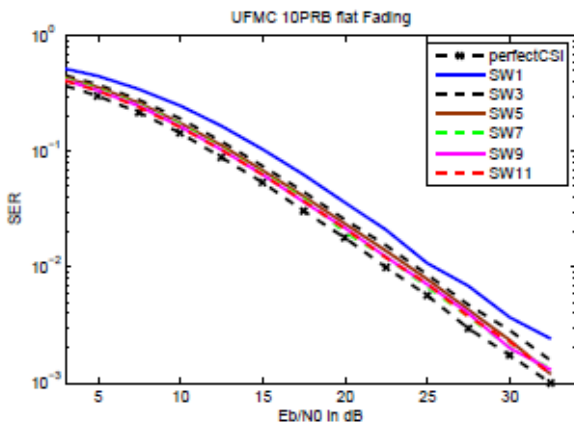


Fig.1.: Single user performance for sliding window UFMC channel estimator.

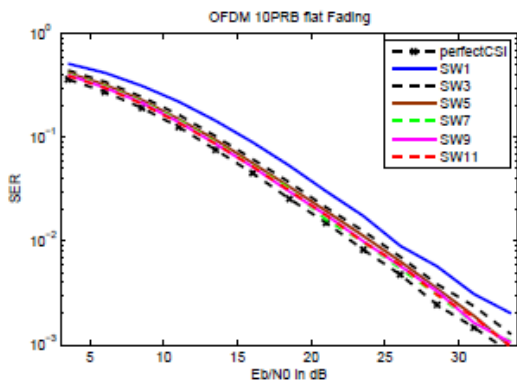


Fig.2.: Single user performance for sliding window OFDM channel estimator

III. MULTIUSER CHANNEL ESTIMATION

A general received signal, if all users are devoted with N_{Alloc} subcarriers, in OFDM is given by

$$Y_O = \mathcal{S}_O^H H + W_O \quad (4)$$

where $H = [H_0, H_1, \dots, H_{N_{user}-1}]$ is channel information vector of size $N_{user} \times N_{Alloc} \times 1$ in the respective subcarrier position for all active users. W_O is noise vector of size $N_{Alloc} \times 1$ and

$$\begin{bmatrix} S_{0,0} & 0 & L & 0 & L & S_{N_{user}-1,0} & 0 & L & 0 \\ 0 & S_{0,1} & L & 0 & L & 0 & S_{N_{user}-1,1} & L & 0 \\ M & M & O & M & L & M & L & O & M \\ 0 & 0 & L & S_{0,N_{Alloc}-1} & L & 0 & 0 & L & S_{N_{user}-1,N_{Alloc}-1} \end{bmatrix} \quad (5)$$

Similarly, the UFMC the received signal can be expressed as

$$Y_U = \mathcal{S}_U^H H + W_U \quad (6)$$

Filter shape is also considered in \mathcal{S}_U^H with

$$\mathcal{S}_U^H = F^H \circ \mathcal{S}_O^H \quad (7)$$

where \circ stands for Hadamard-product of matrices.

Mathematically, final channel estimates for multiuser systems for OFDM and UFMC systems taking sliding window algorithm into account may be written as

$$\hat{H} = S^+ Y_{O,U} = H + S^+ W_{O,U} \quad (8)$$

where $S^+ (= S^H (SS^H)^{-1})$ is Moore-Penrose inverse of S for both UFMC and OFDM systems and

$$S = \begin{bmatrix} S_{0,i} & L & S_{N_{user}-1,i} \\ S_{0,i+1} & L & S_{N_{user}-1,i+1} \\ M & O & M \\ S_{0,i+L_{SW}-1} & L & S_{N_{user}-1,i+L_{SW}-1} \end{bmatrix} \quad (9)$$

for OFDM

and

$$S = \begin{bmatrix} F_{0,i} S_{0,i} & L & F_{N_{user}-1,i} S_{N_{user}-1,i} \\ F_{0,i+1} S_{0,i+1} & L & F_{N_{user}-1,i+1} S_{N_{user}-1,i+1} \\ M & O & M \\ F_{0,i+L_{SW}-1} S_{0,i+L_{SW}-1} & L & F_{N_{user}-1,i+L_{SW}-1} S_{N_{user}-1,i+L_{SW}-1} \end{bmatrix} \quad (10)$$

for UFMC



Combination of DFT-sequence and sliding window, the performance of single and two UEs are examined for frequency selective channel without considering edge subcarriers in figure 3. For a sliding window size of 11, the MSE in UFMC systems are 0.7dB larger than that in OFDM systems.

IV. RESULTS

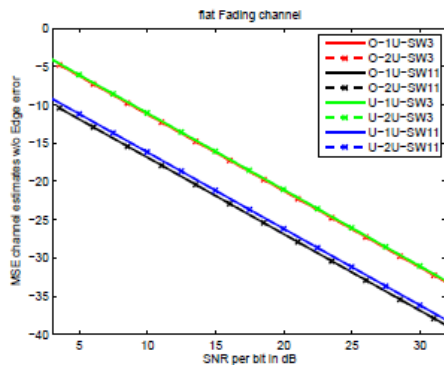


Fig. 3: Performance of Multiuser channel estimator for frequency selective channel without considering edge subcarriers for DFT-sequences.

Next, Simulations are carried out and the results are shown in Fig. 4 for the case of flat-fading channels for OFDM and UFMC systems with different edge subcarrier treatments and sliding window size of 11.

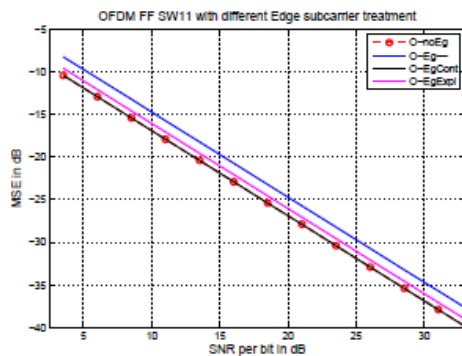


Fig. 4: Comparison of various edge subcarrier techniques in OFDM frequency selective channel model.

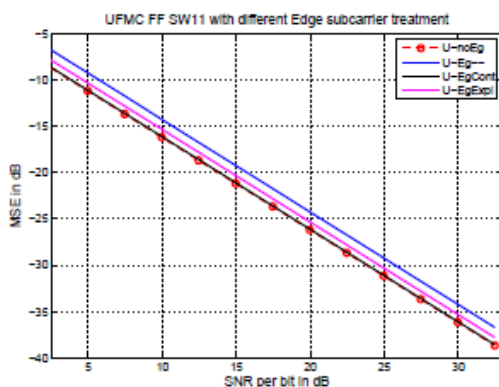


Fig. 5: Comparison of various edge subcarrier techniques in UFMC frequency selective channel model.

Next, the performance among various pilots for flat-fading [6] channel model is compared for a window size of 3 in Fig.5. Simulation results show that orthogonal DFT-sequence has better performance because of its orthogonality.

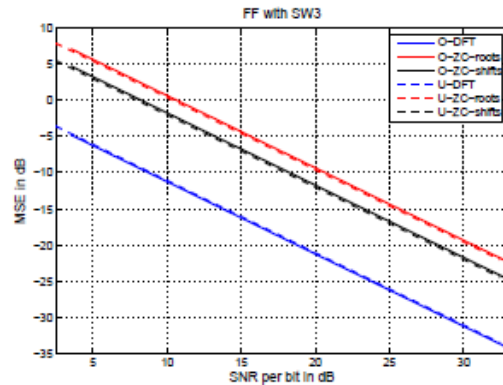


Fig.6: Performance comparison of various pilots in frequency selective channel model.

V. CONCLUSIONS

In this paper, synchronization (pilot-based and cyclic prefix-based) is analyzed and investigated. From simulation results, it is observed that pilot-based synchronization method outperforms cyclic prefix-based synchronization with little added in computational complexity. Further, channel estimation along with pilot sequences are studied and investigated for OFDM and UFMC systems. Simulation results are provided for performance comparison of OFDM and UFMC systems in flat fading channel model. These results show that better performance gains were achieved for UFMC systems for the investigated flat-fading channel model.

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